

## Systems

UFC/DC  
SA (CK0191)  
2018.1

### General concepts

Modelling,  
identification and  
analysis

Control,  
optimisation and  
validation

Fault diagnosis

### Classification

#### Representation

Input-output  
representation

State-space  
representation

The model

#### Examples

#### System properties

Dynamical v  
Instantaneous

Linear v Nonlinear

Stationary v  
non-stationary

Proper v improper

With v without  
delay

## Systems Stochastic algorithms

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### A system

A **system** can be defined as a set of elements (or components) that cooperate to perform a specific functionality that would be otherwise impossible to perform for the individual components alone

This definition is very fine, but it does not highlight an important fact

- The dynamical behaviour of the system

For us a central paradigm is that a system is subjected to external stimuli

- Stimuli influence the temporal evolution of the system itself

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## Systems (cont.)

### A system (reloaded)

A system is a physical entity, typically consisting of different interacting components, that responds to external stimuli producing a *determined/specific* dynamical behaviour

## Systems (cont.)

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We study how to mathematically model a broad variety of systems

Our scope is to analyse their dynamical behaviour

- ↪ We want to operate them appropriately
- ↪ The design of control devices
- ↪ Under external stimuli

The methodological approach shall be formal and system independent

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# General concepts

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There is a wide spectrum of problems that spin around systems theory

↪ System modelling, identification and analysis

- System control, optimisation and verification

- System diagnosis

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# Modelling, identification and analysis

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To study a system, the availability of a **mathematical model** is crucial

↪ A quantitative description of the behaviour of the system

The model is often constructed on the knowledge of the component devices

- A knowledge of the laws the system obeys to must be available, too

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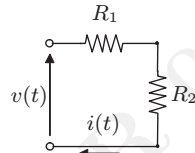
With v without  
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## Modelling (cont.)

### Example

Consider the electric circuit consisting of two serially arranged resistors

The current flow  $i(t)$  [A] thru the system depends on tension  $v(t)$  [V]



- $R_1 = 1 \text{ } [\Omega]$

- $R_2 = 3 \text{ } [\Omega]$

Both resistors can be assumed to follow Ohm's law<sup>1</sup>

$$\rightsquigarrow v(t) = (R_1 + R_2)i(t) = 4i(t)$$

<sup>1</sup>The potential difference ('voltage') across an ideal conductor is proportional to the current that flows through it. The proportionality constant is called 'resistance'.

## Identification

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At times, we only have an incomplete knowledge on the system's devices

- The model must be constructed from observations
- (Observations of the system behaviour)

**Case A)** We have a knowledge on the type/number of component devices

- Not all of their parameters are known
- System observations are available

$\rightsquigarrow$  **Parametric identification**

$\rightsquigarrow$  **White-box identification**

**Case B)** We have no knowledge on the components and their parameters

- Observations of the system are available

$\rightsquigarrow$  **Black-box identification**

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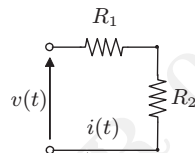
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## Identification (cont.)

### Example

Consider the electric circuit consisting of two serially arranged resistors

The current flow  $i(t)$  [A] thru the system depends on tension  $v(t)$  [V]



- $R_1 = ?$

- $R_2 = ?$

Both resistors can still be assumed to follow Ohm's laws,

$$v(t) = (R_1 + R_2)i(t) = Ri(t)$$

$R$  is now an unknown system parameter

$\rightsquigarrow$  It can be identified from data

## Identification (cont.)

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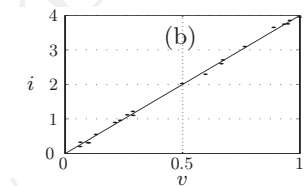
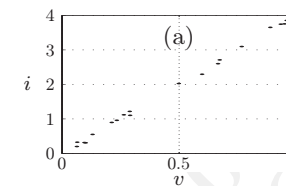
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We can observe the system by collecting  $N$  pairs of measurements  $\{(v_k, i_k)\}_{k=1}^N$



Often, such points will not be perfectly aligned along a line with slope  $R$

$\rightsquigarrow$  **Disturbances** alter the behaviour to the system

$\rightsquigarrow$  **Measurement errors** are always present

We choose  $R$  corresponding to the line that *best* approximates the data

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## Analysis

**Systems analysis** is about forecasting the future behaviour of a system

↪ Based on the external stimuli it is subjected to

The availability of a mathematical model of the system is fundamental

- Needed to approach the problem in a quantitative manner

## Analysis (cont.)

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## Example

The marine ecosystem is described thru time evolution of its fauna and flora

- Birth-growth-dead processes

The behaviour of the system is influenced by many factors

- Climate conditions, availability of food, ...
- Human predators, pollutants, ...
- ..., and so on

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## Analysis (cont.)

They *recently* spoke of reducing  $CO_2$  emissions by injecting it into the sea

- $CO_2$  dissolves in sea water

The lack of a valid model limits our understanding about the system

- We do not know the response of the ecosystem

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**Control, optimisation and  
validation**  
General concepts

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## Control

The objective of **control** is to impose a desired behaviour on a system

We need to explicitly formulate what we mean by 'desired behaviour'

- ↪ The **specifications** that such behaviour must satisfy

We need to design a device for implementing this task, the **controller**

- ↪ The scope of a controller is to stimulate the system
- ↪ Drive its evolution toward the desired behaviour

## Control (cont.)

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### Example

Consider a conventional network for the distribution of drinking water

- Water pressure must be kept constant throughout the net

We can measure pressure at various network locations

- Locations have nominal (target) pressure values

Specs suggest instantaneous variations within  $\pm 10\%$  of nominal value

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## Control (cont.)

We identify two stimuli that act on the system (and modify its behaviour)

- The flow-rate of water withdrawn from the network
- The pressure imposed by the network pumps

We cannot control water withdrawals, they are understood as disturbances

Pump pressures can be manipulated to meet specifications

- This manipulation is done by the controller



## Optimisation

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Achieve a certain system's behaviour, while optimising a performance index

- **Optimisation** can be understood as a special case of control

We impose a desired behaviour, while optimising a **performance index**

- The index measures the quality of the behaviour of the system
- (Economic or operational terms)

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## Optimisation (cont.)

### Example

Consider a conventional suspension system of a conventional car

It is designed to satisfy two different needs

- ~> Appropriate passengers' comfort
- ~> Good handling in all conditions

Modern cars have suspensions based on 'semi-active' technology

- A controller (dynamically) changes the dumping factor
- It guarantees (a compromise between) the two needs

The optimiser takes into account of cabin and wheel oscillations



## Validation

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Suppose that a mathematical model of a system under study is available

- Suppose that a set of desired properties can be formally expressed

**Validation** allows to see whether the model satisfies the properties

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## Validation (cont.)

### Example

Consider a conventional elevator

The system is controlled to guarantee that it responds correctly to requests

- The controller is an abstract machine
- Programmable logic controller (PLC)

Formal verification can be used to guarantee the correct functioning



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## Fault diagnosis

Systems deviate from nominal behaviour because of occurrence of faults

- ↪ We need to detect the presence of an anomaly
- ↪ We need to identify the typology of fault
- ↪ We need to devise a corrective action

### Fault diagnosis

## Fault diagnosis (cont.)

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### Example

The human body is a complex system subjected to many potential faults

- We conventionally call them diseases

Consider the presence of fever, or another anomalous condition

- Symptoms reveal the presence of a disease

A doctor, once identified the pathology, prescribes a therapy

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## Classification

The diversity of systems leads to a number of methodological approaches

- Each approach pertains a particular class of models

Conventional methodological approaches and model/system classification

### By typology

- ↪ Time-evolving systems
- Discrete-event systems
- Hybrid systems

### By representation

- Input-output models
- ↪ State-space models

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## Time-evolving systems

### Time-evolving systems

The system/model behaviour is described with functions, or signals

- The independent variable is time ( $t$  or  $k$ )

#### ↪ Continuous time-evolving systems

- The time variable varies continuously

#### ↪ Discrete time-evolving systems

- The time variable takes discrete values

The signal can take values in a discrete set, **digital time-evolving systems**

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## Time-evolving systems

### Systems by typology

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## Time-evolving systems (cont.)

The evolution of such systems is completely based on the passage of time

The functions of system behaviour satisfy differential/difference equations

- A specification of the relation between functions and their changes

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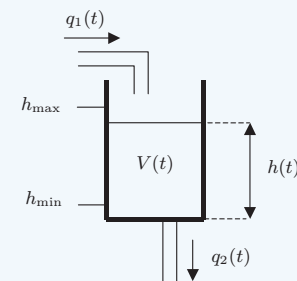
## Time-evolving systems(cont.)

### Example

#### Continuous time-evolving systems

Consider a tank whose volume of liquid  $V(t)$  [ $\text{m}^3$ ] varies in time

- Variation is due to input and output flow rates
- (By two externally operated pumps)



Tank cannot be emptied or filled

$$\leadsto \frac{d}{dt}V(t) = q_1(t) - q_2(t)$$

- Output flow  $q_2(t) \geq 0$  [ $\text{m}^3\text{s}^{-1}$ ]
- Input flow  $q_1(t) \geq 0$  [ $\text{m}^3\text{s}^{-1}$ ]

The differential equation is a relation between continuous-time functions

$$V(t), q_1(t), q_2(t)$$



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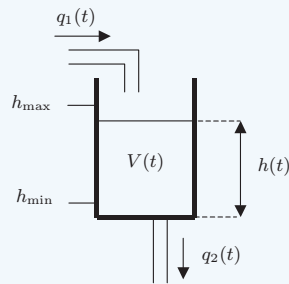
### Example

#### Discrete time-evolving systems

Consider a tank whose volume of liquid  $V(t)$  [m<sup>3</sup>] varies in time

- Measurements are not continuously available
- Acquisitions only every  $\Delta t$  units of time

We are interested in the system behaviour at  $0, \Delta t, 2\Delta t, \dots, k\Delta t, \dots$



We consider discrete-time variables

- $V(k) = V(k\Delta t)$
- $q_1(k) = q_1(k\Delta t)$
- $q_2(k) = q_2(k\Delta t)$

For  $k = 0, 1, 2, \dots$

## Time-evolving systems(cont.)

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We approximate the derivative with the difference quotient

$$\frac{d}{dt}V(t) \approx \frac{\Delta V}{\Delta t} = \frac{V(k+1) - V(k)}{\Delta t}$$

Multiplying both sides of  $\frac{\Delta V}{\Delta t} = q_1(k) - q_2(k)$  by  $\Delta t$  yields

$$\leadsto V(k+1) - V(k) = q_1(k)\Delta t - q_2(k)\Delta t$$

The difference equation is a relation between discrete-time functions

$$V(k), q_1(k), q_2(k)$$



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## Discrete-event systems

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## Discrete-event systems

### Discrete-event systems

Dynamic systems whose *states* take logical or symbolic values (not numeric)

The behaviour is characterised by the occurrence of instantaneous events

$\leadsto$  [At irregular (perhaps unknown) times]

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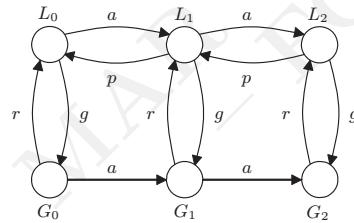
## Discrete-event systems (cont.)

### Example

#### Discrete-event systems

Consider a depot where parts are awaiting to be processed by some machine

- The number of parts awaiting to be processed cannot be larger than 2
- The machine can be either healthy (working) or faulty (stopped)



The state of the system

- Number of awaiting parts
- Status of the machine

The events of the system

- Changes in state

## Discrete-event systems (cont.)

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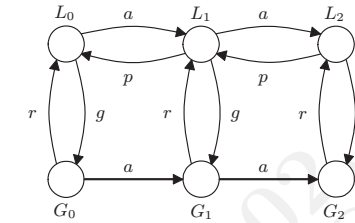
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delay



Six possible states

- $L_0, L_1$  and  $L_2$
- $G_0, G_1$  and  $G_2$

- $L_0$ , the machine is working and the depot is empty
- $L_1$ , the machine is working and there is one part in the depot
- $L_2$ , the machine is working and there are two parts in the depot
- $G_0$ , the machine is not working and the depot is empty
- $G_1$ , the machine is not working and there is one part in the depot
- $G_2$ , the machine is not working and there are two parts in the depot

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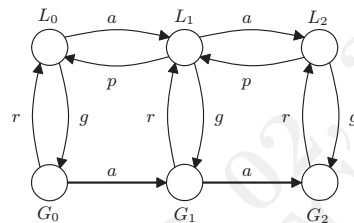
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delay

## Discrete-event systems (cont.)



Four possible events

- $a$  and  $p$
- $g$  and  $r$

- $a$ , a new part arrives to the depot
- $p$ , the machines takes one part from the depot
- $g$ , the machine gets faulty
- $r$ , the machine gets fixed

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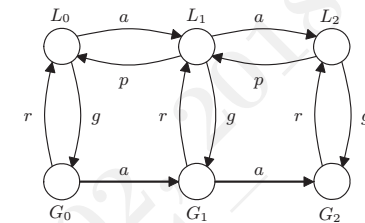
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## Discrete-event systems (cont.)



Event  $a$  can only occur when the depot does not have two parts

$$a \rightsquigarrow \begin{cases} L_i \rightarrow L_{i+1} \\ G_i \rightarrow G_{i+1} \end{cases}$$

Event  $p$  can only occur when the depot is not empty

$$p \rightsquigarrow \begin{cases} L_i \rightarrow L_{i-1} \end{cases}$$

Event  $g$  and  $r$  determine the switches  $L_i \rightarrow G_i$  and  $G_i \rightarrow L_i$ , respectively

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# Hybrid systems

## Systems by typology

## Hybrid systems

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A model that combines time-evolving dynamics and discrete-event dynamics

~> Thus, they are the most general class of dynamical systems

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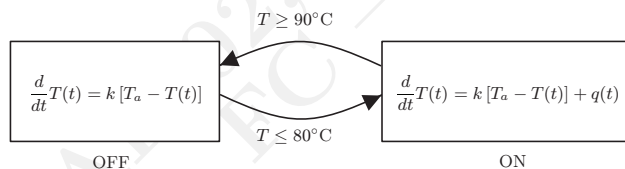
## Hybrid systems (cont.)

### Example

#### Hybrid systems

Consider a modern sauna, a cabin where the temperature is regulated

- A thermostat controls a stove used as heat generator
- Keep the temperature between 80°C and 90°C



The thermostat can be represented using a discrete-event model

- Switch the heater ON
- Switch the heater OFF

The cabin can be represented using a time-evolving model

- Temperature  $T(t)$

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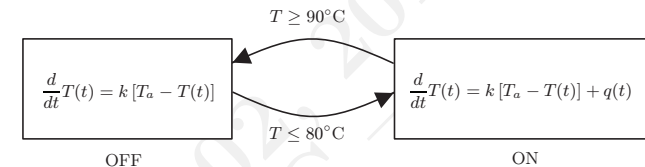
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## Hybrid systems(cont.)



When the thermostat is OFF

$T(t)$  in the cabin decreases, heat exchanged with the outside [ $T_a < T(t)$ ]

$$\frac{d}{dt}T(t) = k[T_a - T(t)], \quad \text{with } k > 0$$

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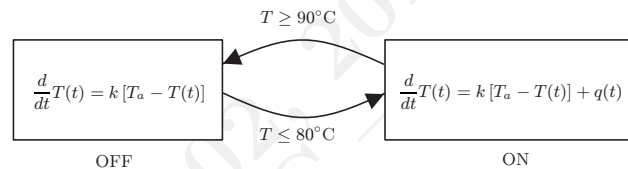
non-stationary

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## Hybrid systems(cont.)



When the thermostat is ON

$T(t)$  in the cabin increases, heat generated by the stove  $q(t)$

$$\frac{d}{dt}T(t) = k[T_a - T(t)] + q(t)$$

## Hybrid systems(cont.)

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The state of the system is  $x = (l, T)$

- A logical variable  $l \in \{\text{ON}, \text{OFF}\}$ , representing the discrete state
- A real function  $T(t) \in \mathcal{R}$ , representing the continuous state



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# Representation Systems

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## Representation

We provide fundamental concepts for the analysis of time-evolving systems

- Evolution generates from the passing of time
- Focus on continuous-time systems

We introduce the two main forms that are used for describing such systems

↪ Mathematic formulations and example(s)

We conclude with a classification, based on some system/model properties

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## Representation (cont.)

Fundamental step to use formal techniques to study time-evolving systems

↪ We describe the system behaviour in terms of functions

There are two possible such model/system descriptions

- **Input-output (IO)** representation

↪ **State-space (SS)** representation

We define the mathematical elements and properties of these representations

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## Input-output representation Representation

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## Input-output representation

The quantities involved in the input-output (IO) representation of a system

### Causes

- Quantities that are generated outside the system
- Their evolution influences the system behaviour
- They are not influenced by the system behaviour

### Effects

- Quantities whose behaviour is influenced by the causes
- Their evolution is influenced by nature of the system

By convention,

$$\rightsquigarrow \left\{ \begin{array}{ll} \text{Causes} & \rightsquigarrow \text{Inputs} \\ \text{Effects} & \rightsquigarrow \text{Outputs} \end{array} \right.$$

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## Input-output representation (cont.)

### A system

The system  $S$  can be seen as an operator or a processing/computing unit

- It assigns a specific evolution to the output variables
- One for each possible evolution of the input variables

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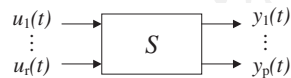
With v without  
delay

## Input-output representation (cont.)

A system can have more than one ( $r$ ) input and more than one ( $p$ ) output

- Both inputs and outputs are assumed to be measurable/observable

A graphical IO system representation



$r$  **inputs** (in  $\mathcal{R}^r$ )

$$\mathbf{u}(t) = [u_1(t) \cdots u_r(t)]^T$$

$p$  **outputs** (in  $\mathcal{R}^p$ )

$$\mathbf{y}(t) = [y_1(t) \cdots y_p(t)]^T$$

### Manipulable inputs

- They can be used for control

### Non-manipulable inputs

- They are called disturbances

## Input-output representation (cont.)

### Example

#### A car (IO representation)

Consider a car also

Let its position and speed be the output variables

- They are both measurable

As input variables, we can consider wheel and gas position

- They are both measurable and manipulable

By acting on the input variables, we influence the output behaviour

- The changes depend on the specific system (car)
- (More precisely, by its dynamics)

Wind speed could be considered as an additional input variable

- It may be measurable but it is hardly manipulable

$r = 3$  inputs and  $p = 2$  outputs (A MIMO system)

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## Input-output representation (cont.)

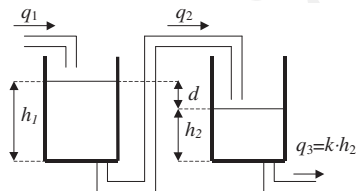
### Example

#### Two tanks (IO representation)

Consider a system consisting of two cylindric tanks, both of base  $B$  [m<sup>2</sup>]

- Output flow-rate from tank 1 is the input flow-rate to tank 2

$\leadsto q_2$



First tank

- Input flow-rate  $q_1$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_1$  is the liquid level [m]

Second tank

- Input flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_3$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_2$  is the liquid level [m]

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## Input-output representation (cont.)

Suppose that flow-rates  $q_1$  and  $q_2$  can be set to some desired value (pumps)

Also suppose that  $q_3$  depends linearly on the liquid level in the tank

- $q_3 = kh_2$  [m<sup>3</sup>s<sup>-1</sup>], with  $k$  [m<sup>2</sup>s<sup>-1</sup>]
- $k$  appropriate constant

### Inputs, $q_1$ and $q_2$

- $\leadsto$  Measurable and manipulable
- $\leadsto$  They influence the liquid levels in the tanks

### Output, $d = h_1 - h_2$

- $\leadsto$  Measurable, but not manipulable
- $\leadsto$  It is influenced indirectly only through inputs

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# State-space representation

## Representation

## State-space representation

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For a given behaviour of the inputs,  $S$  defines the behaviour of the outputs

- ↪ The output at time  $t$  is not only dependent on the inputs at time  $t$
- ↪ It also depends on the past behaviour (evolution) of the system

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## State-space description(cont.)

### Example

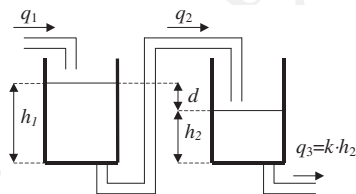
#### Two tanks (SS representation)

Consider a system consisting of two cylindric tanks, both of base  $B$  [m<sup>2</sup>]

Let  $d_0 = h_{1,0} - h_{2,0}$  be the output variable at time  $t_0$

- ( $h_{1,0}$  and  $h_{2,0}$  are the liquid levels at time  $t_0$ )

Suppose that all input variables ( $q_1$  and  $q_2$ ) are zero at time  $t_0$



- $q_{1,0} = 0$
- $q_{2,0} = 0$

Output  $d(t)$  at any time  $t > t_0$  depends on input values  $q_1(t)$  and  $q_2(t)$

- Over the entire interval  $[t_0, t]$



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## State-space representation (cont.)

We can take this into account by introducing an *intermediate* variable

A variable that *exists* between inputs and outputs

- The **state** variable of the system

$r$  **inputs** (in  $\mathcal{R}^r$ )

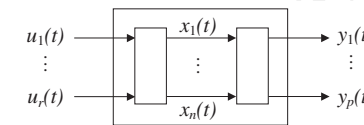
$$\mathbf{u}(t) = [u_1(t) \cdots u_r(t)]^T$$

$n$  **states** (in  $\mathcal{R}^n$ )

$$\mathbf{x}(t) = [x_1(t) \cdots x_n(t)]^T$$

$p$  **outputs** (in  $\mathcal{R}^p$ )

$$\mathbf{y}(t) = [y_1(t) \cdots y_p(t)]^T$$



The state condenses information about past and present of the system

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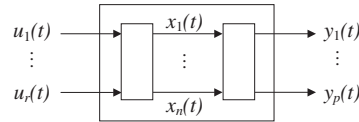
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delay

## State-space representation (cont.)



The **state vector**  $\mathbf{x}(t) = [x_1(t) \cdots x_n(t)]^T$  has  $n$  components

↪ We say that  $n$  is the order of the system

- (In this representation)

## State-space representation (cont.)

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### Definition

#### State variable

The **state** of a system at time  $t_0$  is a variable that contains the necessary information to univocally determine the behaviour of output  $\mathbf{y}(t)$  for  $t \geq t_0$

- Given the behaviour of input  $\mathbf{u}(t)$  for  $t \geq t_0$  and the state itself at  $t_0$

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## State-space description (cont.)

In general, it is possible to select different physical entities as state variables

- The state variable is neither univocally defined, nor it is determined
- It is anything that can be seen as an *internal cause* of evolution
- (In general)

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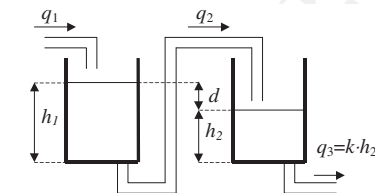
With v without  
delay

## State-space representation (cont.)

### Example

#### Two tanks (SS representation)

Consider a system consisting of two cylindric tanks, both of base  $B$  [m<sup>2</sup>]



First tank

- Input flow-rate  $q_1$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_1$  is the liquid level [m]

Second tank

- Input flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_3$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_2$  is the liquid level [m]

Let  $d_0 = h_{1,0} - h_{2,0}$  be the output variable at time  $t_0$

- $h_{1,0}$  and  $h_{2,0}$  are the liquid levels at time  $t_0$

As state variable, select the volume of liquid in the tanks,  $V_1(t)$  and  $V_2(t)$



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## State-space representation (cont.)

We shall see that we are able to evaluate the output  $d(t)$  for  $t > t_0$

- ↪ Need to know the initial state,  $V_{1,0}$  and  $V_{2,0}$ , at  $t_0$
- ↪ Need to know the input,  $q_1(t)$  and  $q_2(t)$ , in  $[t_0, t]$



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## State-space representation (cont.)

Common to choose as state those variables that characterise system energy

- For a cylindric tank of base  $B$  and liquid level  $h(t)$ , the potential energy at time  $t$  is  $E_p(t) = 1/2\rho g V^2(t)/B$ , with  $\rho$  the density of the liquid and  $V(t) = Bh(t)$ .  $V(t)$  or equivalently  $h(t)$  can be used as state variable
- For a spring with elastic constant  $k$ , the potential energy at time  $t$  is  $E_k(t) = 1/2kz^2(t)$  with  $z(t)$  the spring deformation with respect to an equilibrium position.  $z(t)$  can be used as state variable
- For a mass  $m$  moving with speed  $v(t)$  on a plane, the kinetic energy at time  $t$  is  $E_m(t) = 1/2mv^2(t)$ .  $v(t)$  can be used as state of the system

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## State-space description(cont.)

Consider a system in which there is energy stored, its state is not zero

- The system can evolve even in the absence of external inputs

The state can be understood as a possible (internal) cause of evolution

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## State-space representation (cont.)

### Example

#### Two tanks (SS representation, reloaded)

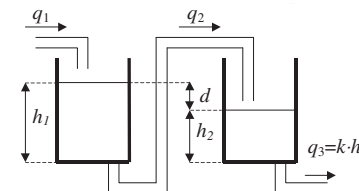
Consider any of the tanks in the two cylindric tanks system, base  $B$  [m<sup>2</sup>]

First tank

- Input flow-rate  $q_1$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_1$  is the liquid level [m]

Second tank

- Input flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_3$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_2$  is the liquid level [m]



Each of the tanks can store a certain amount of potential energy

- The amount depends on the liquid volumes (levels)

The entire (two-tank) system has order 2



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# Mathematical model

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System analysis studies the relations between the system inputs and outputs (IO) or, alternatively, between the system inputs, states and outputs (SS)

We are given certain input functions

↪ Interest in understanding how states and outputs evolve in time

We need a model to describe quantitatively the system behaviour

- The relations between inputs (states) and outputs

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## Mathematical model (cont.)

### Input-output model

The relationship between the system output  $\mathbf{y}(t) \in \mathcal{R}^p$  and its derivatives, the system input  $\mathbf{u}(t) \in \mathcal{R}^r$  and its derivatives (a **differential equation**)

### State-space model

It describes how the evolution  $\dot{\mathbf{x}}(t) \in \mathcal{R}^n$  of the system state depends on the state  $\mathbf{x}(t) \in \mathcal{R}^n$  itself and on the output  $\mathbf{u}(t) \in \mathcal{R}^r$  (the **state equation**)

It describes how the system output  $\mathbf{y}(t) \in \mathcal{R}^p$  depends on system state  $\mathbf{x}(t) \in \mathcal{R}^n$  and on system input  $\mathbf{u}(t) \in \mathcal{R}^r$  (the **output transformation**)

# Input-output model

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## Input-output model

The IO model of a SISO system is given as a differential equation

$$h \left[ \underbrace{y(t), \dot{y}(t), \dots, y^{(n)}(t)}_{\text{output}}, \underbrace{u(t), \dot{u}(t), \dots, u^{(m)}(t)}_{\text{input}}, \underbrace{t}_{\text{time}} \right] = 0$$

- $\dot{y}(t) = \frac{d}{dt}y(t), \dots$  and  $\dots, y^{(n)}(t) = \frac{d^n}{dt^n}y(t)$
- $\dot{u}(t) = \frac{d}{dt}u(t), \dots$  and  $\dots, u^{(m)}(t) = \frac{d^m}{dt^m}u(t)$

$h$  is a multi-parametric function that depends on the system

- $n$  is the maximum order of derivation of the output
- $m$  is the maximum order of derivation of the input

The **order of the system (model)** is  $n$

## Input-output model (cont.)

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### Example

Consider a system model given by the differential equation

$$\overbrace{2 \dot{y}(t)y(t)}^{\text{output}} + 2 \sqrt{\overbrace{t}^{\text{time}} \underbrace{u(t)\ddot{u}(t)}_{\text{input}}} = 0$$

We have,

- Output order of derivation,  $n = 1$
- Input order of derivation,  $m = 2$

Function  $h$  links  $y$  and  $\dot{y}$ , and  $u$  and  $\ddot{u}$ , and  $t$  is the independent variable

The relationship *explicitly* depends on the independent variable (time)

$$\rightsquigarrow \sqrt{t}$$

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## Input-output model (cont.)

The IO model of a MIMO system with  $p$  outputs and  $r$  inputs

$$\begin{cases} h_1 \left[ \underbrace{y_1(t), \dot{y}_1(t), \dots, y_1^{(n_1)}(t)}_{\text{output 1}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{1,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{1,r})}(t)}_{\text{input r}} \right] \\ = 0 \\ h_2 \left[ \underbrace{y_2(t), \dot{y}_2(t), \dots, y_2^{(n_2)}(t)}_{\text{output 2}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{2,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{2,r})}(t)}_{\text{input r}} \right] \\ = 0 \\ \vdots \\ h_p \left[ \underbrace{y_p(t), \dot{y}_p(t), \dots, y_p^{(n_p)}(t)}_{\text{output p}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{p,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{p,r})}(t)}_{\text{input r}} \right] \\ = 0 \end{cases}$$

$h_i$  ( $i = 1, \dots, p$ ) are multi-parametric functions depending on the system

- $n_i$ , max order of derivation of the  $i$ -th component of output  $y_i(t)$
- $m_i$ , max order of derivation of the  $i$ -th component of input  $u_i(t)$

A total of  $p$  differential equations

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## State-space model

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## State-space model

The SS model of a SISO system is not a differential equation of order  $n$

## State-space model (cont.)

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### State equation

$$\begin{cases} \dot{x}_1(t) = f_1[x_1(t), \dots, x_n(t), u(t), t] \\ \dot{x}_2(t) = f_2[x_1(t), \dots, x_n(t), u(t), t] \\ \vdots \\ \dot{x}_n(t) = f_n[x_1(t), \dots, x_n(t), u(t), t] \end{cases}$$

It links the derivative of each state with the other states and the input

### Output transformation

$$y(t) = g[x_1(t), \dots, x_n(t), u(t), t]$$

It further links the output with each state variable and the input

$f_i$  with  $i = 1, \dots, n$  and  $g$  are multi-parametric functions

- They depend on (are) the dynamics of the system

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## State-space model (cont.)

Let  $\dot{\mathbf{x}}(t)$  be the vector whose components are the derivatives of the state

$$\dot{\mathbf{x}}(t) = \frac{d}{dt}\mathbf{x}(t) = \begin{bmatrix} \dot{x}_1(t) \\ \vdots \\ \dot{x}_n(t) \end{bmatrix} = \begin{bmatrix} \frac{d}{dt}x_1(t) \\ \vdots \\ \frac{d}{dt}x_n(t) \end{bmatrix}$$

## State-space model (cont.)

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### State-space model

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), u(t), t] \\ y(t) = g[\mathbf{x}(t), u(t), t] \end{cases}$$

$\mathbf{f}$  is a vectorial function whose  $i$ -th component is  $f_i$ , with  $i = 1, \dots, n$

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## State-space model (cont.)

The SS model of a MIMO system with  $r$  inputs and  $p$  outputs

### State equation

$$\begin{cases} \dot{x}_1(t) = f_1[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \dot{x}_2(t) = f_2[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \vdots \\ \dot{x}_n(t) = f_n[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \end{cases}$$

### Output transformation

$$\begin{cases} y_1(t) = g_1[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ y_2(t) = g_2[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \vdots \\ y_p(t) = g_p[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \end{cases}$$

Multi-parametric functions depending on the system

- $f_i$  with  $i = 1, \dots, n$
- $g_i$  with  $i = 1, \dots, p$

## State-space model (cont.)

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### State-space model

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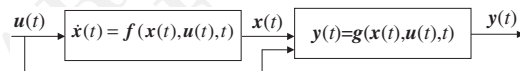
## State-space model (cont.)

The state equation is a set of  $n$  first-order differential equations

- Regardless of the fact that the system is SISO or MIMO

The output transformation is a scalar or vectorial algebraic equation

- Depending on the number  $p$  of output variables



## State-space model (cont.)

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The state-space representation of a system is central in our methods

- It offers a consistent framework for analysing systems
- Analysis of systems of arbitrary degree of complexity

Conversion of a scalar  $n$ -th order ordinary differential equation

- $\rightsquigarrow n$  first-order ordinary differential equations
- $\rightsquigarrow$  A first-order vector equation, dimension  $n$

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## State-space model (cont.)

### Example

Consider the third-order scalar equation

$$\ddot{x}(t) + c_3\ddot{x}(t) + c_2\dot{x}(t) + c_1x(t) = bu(t)$$

We define,

$$\mathbf{x}(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix} = \begin{bmatrix} x(t) \\ \dot{x}(t) \\ \ddot{x}(t) \end{bmatrix}$$

Thus,

$$\dot{\mathbf{x}}(t) = \begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \\ \dot{x}_3(t) \end{bmatrix} = \begin{bmatrix} x_2(t) \\ x_3(t) \\ -c_3x_3(t) - c_2x_2(t) - c_1x_1(t) + bu(t) \end{bmatrix}$$

This equation can be first integrated to get  $x_3(t) [\rightsquigarrow \ddot{x}(t)]$

- Two more integrations to get  $x_2(t) [\rightsquigarrow \dot{x}(t)]$  and  $x_1 [\rightsquigarrow x(t)]$

## State-space model (cont.)

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The three scalar differential equations that must be solved

$$\dot{\mathbf{x}}(t) = \begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \\ \dot{x}_3(t) \end{bmatrix} = \begin{bmatrix} x_2(t) \\ x_3(t) \\ [-c_3x_3(t) - c_2x_2(t) - c_1x_1(t) + bu(t)] \end{bmatrix}$$

$$= \begin{bmatrix} f_1[\mathbf{x}(t), u(t), t] \\ f_2[\mathbf{x}(t), u(t), t] \\ f_3[\mathbf{x}(t), u(t), t] \end{bmatrix}$$

A single vector (state) equation

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## State-space model (cont.)

### Example

Consider two interconnected second-order systems

$$\ddot{y} - a_2(\dot{z} - \dot{y}) - a_1(z - y)^2 = b_1u_1 + b_2u_2$$

$$\ddot{z} - c_2\dot{z}^2 - c_1(y + z) = du_1^2$$

The state vector could be

$$\mathbf{x} = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} = \begin{bmatrix} x_2 \\ a_2(x_4 - x_2) + a_1(x_3 - x_1)^2 = b_1u_1 + b_2u_2 \\ x_4 \\ c_2x_4^2 + c_1(x_3 + x_1) + du_1^2 \end{bmatrix}$$

## State-space model (cont.)

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The state vector is not necessarily unique

$$\mathbf{x} = \begin{bmatrix} y \\ \dot{y} \\ (z - y) \\ (\dot{z} - \dot{y}) \end{bmatrix}$$

The dynamic equation would get adjusted, accordingly

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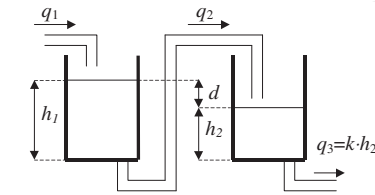
Proper v improper

With v without  
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# Examples Systems

## Examples

### Example



#### First tank

- Input flow-rate  $q_1$  [ $\text{m}^3\text{s}^{-1}$ ]
- Output flow-rate  $q_2$  [ $\text{m}^3\text{s}^{-1}$ ]
- $h_1$  is the liquid level [m]

#### Second tank

- Input flow-rate  $q_2$  [ $\text{m}^3\text{s}^{-1}$ ]
- Output flow-rate  $q_3$  [ $\text{m}^3\text{s}^{-1}$ ]
- $h_2$  is the liquid level [m]

- $u_i = q_i$  with  $i = 1, 2$ , the input variables
- $y = d$ , the output variable
- $x_1 = V_1$  and  $x_2 = V_2$ , the state variables

We are interested in the IO and the SS representations of the system

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## Examples (cont.)

For an incompressible fluid, by mass conservation

$$\begin{cases} \frac{dV_1(t)}{dt} = q_1(t) - q_2(t) \\ \frac{dV_2(t)}{dt} = q_2(t) - q_3(t) = q_2(t) - kh_2(t) \end{cases} \quad (1)$$

We have  $h_1 = V_1/B$ ,  $h_2 = V_2/B$ , and  $q_3 = kh_2$ , thus

$$\begin{cases} \dot{h}_1(t) = \frac{1}{B}q_1(t) - \frac{1}{B}q_2(t) \\ \dot{h}_2(t) = \frac{1}{B}q_2(t) - \frac{1}{B}q_3(t) = \frac{1}{B}q_2(t) - \frac{k}{B}h_2(t) \end{cases}$$

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## Examples (cont.)

Moreover, we have  $y(t) = d(t) = h_1(t) - h_2(t)$

Thus,

$$\begin{aligned} \dot{y}(t) &= \dot{d}(t) = \dot{h}_1(t) - \dot{h}_2(t) = \underbrace{\left[ \frac{1}{B}q_1(t) - \frac{1}{B}q_2(t) \right]}_{\dot{h}_1(t)} - \underbrace{\left[ \frac{1}{B}q_2(t) - \frac{k}{B}h_2(t) \right]}_{\dot{h}_2(t)} \\ &= \frac{1}{B}q_1(t) - \frac{2}{B}q_2(t) + \frac{k}{B}h_2(t) \\ &= \frac{1}{B}u_1(t) - \frac{2}{B}u_2(t) + \frac{k}{B}[h_1(t) - y(t)] \end{aligned}$$

We used  $u_1(t) = q_1(t)$  and  $u_2(t) = q_2(t)$

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## Examples (cont.)

$$\dot{y}(t) = \frac{1}{B}u_1(t) - \frac{2}{B}u_2(t) + \frac{k}{B}h_1(t) - \frac{k}{B}y(t)$$

By taking the second derivative of  $y(t)$ , we have

$$\begin{aligned}\ddot{y}(t) &= \frac{1}{B}\dot{u}_1(t) - \frac{2}{B}\dot{u}_2(t) + \frac{k}{B}\dot{h}_1(t) - \frac{k}{B}\dot{y}(t) \\ &= \frac{1}{B}\dot{u}_1(t) - \frac{2}{B}\dot{u}_2(t) + \underbrace{\frac{k}{B^2}u_1(t) - \frac{k}{B^2}u_2(t)}_{\frac{k}{B}\dot{h}_1(t)} - \frac{k}{B}\dot{y}(t)\end{aligned}$$

The IO system representation is an ordinary differential equation

$$\leadsto \ddot{y}(t) + \frac{k}{B}\dot{y}(t) = \frac{1}{B}\dot{u}_1(t) - \frac{2}{B}\dot{u}_2(t) + \frac{k}{B^2}u_1(t) - \frac{k}{B^2}u_2(t)$$

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## Examples (cont.)

$$\ddot{y}(t) + \frac{k}{B}\dot{y}(t) - \frac{k}{B^2}u_1(t) - \frac{1}{B}\dot{u}_1(t) + \frac{k}{B}u_2(t) + \frac{2}{B}\dot{u}_2(t) = 0$$

The obtained model is in the general IO form

$$\begin{cases} h_1 \left[ \underbrace{y_1(t), \dot{y}_1(t), \dots, y_1^{(n_1)}(t)}_{\text{output 1}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_1,1)}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_1,r)}(t)}_{\text{input r}}, t \right] \\ = 0 \\ \vdots \\ h_p \left[ \underbrace{y_p(t), \dot{y}_p(t), \dots, y_p^{(n_p)}(t)}_{\text{output p}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_p,1)}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_p,r)}(t)}_{\text{input r}}, t \right] \\ = 0 \end{cases}$$

$$\leadsto p = 1, n_1 = 2$$

$$\leadsto r = 2, m_1 = m_2 = 1$$

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## Examples (cont.)

The SS system representation is derived from mass conservation

$$\leadsto \begin{cases} \frac{dV_1(t)}{dt} = q_1(t) - q_2(t) \\ \frac{dV_2(t)}{dt} = q_2(t) - q_3(t) = q_2(t) - kh_2(t) \end{cases}$$

The state equation is obtained by setting  $h_2 = x_2/B$

The output transformation,

$$y(t) = \frac{1}{B}x_1(t) - \frac{1}{B}x_2(t)$$

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## Examples (cont.)

The resulting SS representation of the system

$$\leadsto \begin{cases} \dot{x}_1(t) = u_1(t) - u_2(t) \\ \dot{x}_2(t) = -\frac{k}{B}x_2(t) + u_2(t) \\ \dot{y}(t) = \frac{1}{B}x_1(t) - \frac{1}{B}x_2(t) \end{cases}$$



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## Examples (cont.)

The model is in the general SS form

### State equation

$$\begin{cases} \dot{x}_1(t) = f_1[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \dot{x}_2(t) = f_2[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \vdots \\ \dot{x}_n(t) = f_n[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \end{cases}$$

### Output transformation

$$\begin{cases} y_1(t) = g_1[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ y_2(t) = g_2[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \\ \vdots \\ y_p(t) = g_p[x_1(t), \dots, x_n(t), u_1(t), \dots, u_r(t), t] \end{cases}$$

## Examples (cont.)

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In general, the choice of the states is not unique

- We could have chosen the levels as states
- $x_1 = h_1$  and  $x_2 = h_2$

$$\rightsquigarrow \begin{cases} \dot{x}_1(t) = -Bu_1(t) - Bu_2(t) \\ \dot{x}_2(t) = -kx_2(t) + Bu_2(t) \\ y(t) = x_1(t) - x_2(t) \end{cases}$$

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We discuss a set of fundamental properties of time-evolving models

- Proper v Improper
- Linear v Non-linear
- With v Without delay
- Dynamical v Instantaneous
- Stationary v Non-stationary
- *Lumped v Distributed parameters*

Yet another way of classifying dynamical systems/models

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# System properties

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## Dynamical v Instantaneous

### Definition

#### Instantaneity

A system is said to be **instantaneous** if the value of the output  $\mathbf{y}(t) \in \mathcal{R}^p$  at time  $t$  only depends on the value of the input  $\mathbf{u}(t) \in \mathcal{R}^r$  at time  $t$

A system is said to be **dynamical**, otherwise

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## Dynamical v Instantaneous (cont.)

### Proposition

#### IO representation - SISO

A necessary and sufficient condition for a SISO system to be instantaneous is that the IO relationship is expressed by an equation in the form

$$h[y(t), \dot{y}(t), \dots, y^{(n)}(t), u(t), \dot{u}(t), \dots, u^{(m)}(t), t] = h[y(t), u(t), t] = 0$$

If a SISO system is instantaneous, the IO relation is an algebraic equation

↪ The order of the derivatives of  $y$  and  $u$  is zero ( $n = m = 0$ )

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## Dynamical v Instantaneous (cont.)

This is necessary but not sufficient for the system to be instantaneous

Consider as IO representation of a SISO system a differential equation

↪ Then, the system is certainly dynamical

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## Dynamical v Instantaneous systems (cont.)

### Example

#### Counter-intuition

Consider as IO representation of a SISO system the algebraic equation

$$y(t) = u(t - T), \quad \text{with } T \in \mathcal{R}^+$$

Such a system is not instantaneous, it is dynamical

↪ Finite time delay system

The output  $y(t)$  at time  $t$  does not depend on the input  $u(t)$  at time  $t$

- It depends on the input value  $u(t - T)$ , at a preceding moment



## Dynamical v Instantaneous (cont.)

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### Proposition

#### IO representation - MIMO

A necessary and sufficient condition for a MIMO system with  $r$  inputs and  $p$  outputs to be instantaneous is that the IO representation is expressed by a system of equations in the form

$$\begin{cases} h_1[y_1(t), u_1(t), u_2(t), \dots, u_r(t), t] = 0 \\ h_2[y_2(t), u_1(t), u_2(t), \dots, u_r(t), t] = 0 \\ \vdots \\ h_p[y_p(t), u_1(t), u_2(t), \dots, u_r(t), t] = 0 \end{cases}$$



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## Dynamical v Instantaneous (cont.)

If a MIMO system is instantaneous, then the following conditions are true

- ↪ The order of the derivatives of  $y_i$  is  $n_i = 0$ , for all  $i = 1, \dots, p$
- ↪ The order of the derivatives of  $u_i$  is  $m_{j,i} = 0$ , for all  $j = 1, \dots, p$  and  $i = 1, \dots, r$

The IO relation can be expressed as a system of  $p$  algebraic equations

If any of the IO relations is a differential equation, the system is dynamical

## Dynamical v Instantaneous

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### Proposition

#### SS representation

Consider a model of a system expressed in SS form

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), \mathbf{u}(t), t] \\ \mathbf{y}(t) = \mathbf{g}[\mathbf{x}(t), \mathbf{u}(t), t] \end{cases}$$

A necessary and sufficient condition for a system to be instantaneous is that the SS model is zero order (i.e., there exists no state vector)

$$\rightsquigarrow \mathbf{y}(t) = \mathbf{g}[\mathbf{u}(t), t]$$



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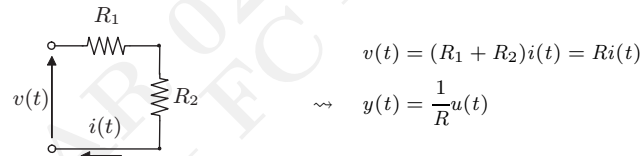
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## Dynamical v Instantaneous

### Example

Consider the two serially arranged resistors, the system is instantaneous

The IO representation corresponds to the SS output transformation



The order of the system is zero (no device to store energy)

## Linear v Nonlinear

### Properties

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## Linear v Nonlinear

### Definition

#### Linearity

A system is said to be **linear** if it obeys the superposition principle

A system is said **nonlinear**, otherwise

## Linear v Non-linear (cont.)

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### Superposition principle

Consider some system

- Let the system response to cause  $c_1$  be equal to effect  $e_1$
- Let the system response to cause  $c_2$  be equal to effect  $e_2$

The system response to cause  $(\alpha c_1 + \beta c_2)$  equals effect  $(\alpha e_1 + \beta e_2)$

- (whatever the constants  $\alpha$  and  $\beta$ )

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## Linear v Non-linear (cont.)

### Proposition

#### IO representation - SISO

A necessary and sufficient condition for a SISO system to be linear is that the IO representation is expressed by a linear differential equation

$$a_0(t)y(t) + a_1(t)\dot{y}(t) + \dots + a_n(t)y^{(n)}(t) = b_0(t)u(t) + b_1(t)\dot{u}(t) + \dots + b_m(t)u^{(m)}(t)$$

The coefficients of the IO representation are, in general, time dependent

## Linear v Non-linear (cont.)

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### Linear differential equation

Consider the differential equation

$$h[y(t), \dot{y}(t), \dots, y^{(n)}(t), u(t), \dot{u}(t), \dots, u^{(m)}(t), t] = 0$$

The equation is linear if and only if the function  $h$  is a linear combination of the output and its derivatives, and of the input and its derivatives

$$\alpha_0(t)y(t) + \alpha_1(t)\dot{y}(t) + \dots + \alpha_n(t)y^{(n)}(t) + \beta_0(t)u(t) + \beta_1(t)\dot{u}(t) + \dots + \beta_m(t)u^{(m)}(t) = 0$$

↪ A zero-sum weighted sum of inputs, outputs, and derivatives

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## Linear v non-linear (cont.)

Consider a MIMO system in IO representation

Each function  $h_i$ ,  $i = 1, \dots, p$ , must be a linear combination of the  $i$ -th component of the output and its  $n_i$  derivatives, and the input and its derivatives

The condition is necessary and sufficient

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## Linear v non-linear (cont.)

### Proposition

#### SS representation

A necessary and sufficient condition for a system to be linear is that state equation and output transformation in the SS model are linear equations

$$\begin{cases} \dot{x}_1(t) = a_{1,1}(t)x_1(t) + \dots + a_{1,n}(t)x_n(t) + b_{1,1}u_1(t) + \dots + b_{1,r}(t)u_r(t) \\ \vdots \\ \dot{x}_n(t) = a_{n,1}(t)x_1(t) + \dots + a_{n,n}(t)x_n(t) + b_{n,1}u_1(t) + \dots + b_{n,r}(t)u_r(t) \\ y_1(t) = c_{1,1}(t)x_1(t) + \dots + c_{1,n}(t)x_n(t) + d_{1,1}u_1(t) + \dots + d_{1,r}(t)u_r(t) \\ \vdots \\ y_p(t) = c_{p,1}(t)x_1(t) + \dots + c_{p,n}(t)x_n(t) + d_{p,1}u_1(t) + \dots + d_{p,r}(t)u_r(t) \end{cases}$$

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## Linear v non-linear (cont.)

$$\rightsquigarrow \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}(t)\mathbf{x}(t) + \mathbf{B}(t)\mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C}(t)\mathbf{x}(t) + \mathbf{D}(t)\mathbf{u}(t) \end{cases} \rightsquigarrow \begin{aligned} \mathbf{A}(t) &= \{a_{i,j}(t)\} \in \mathcal{R}^{n \times n} \\ \mathbf{B}(t) &= \{b_{i,j}(t)\} \in \mathcal{R}^{n \times r} \\ \mathbf{C}(t) &= \{c_{i,j}(t)\} \in \mathcal{R}^{p \times n} \\ \mathbf{D}(t) &= \{d_{i,j}(t)\} \in \mathcal{R}^{p \times r} \end{aligned}$$

Coefficient matrices **A**, **B**, **C** and **D** are, in general, time dependent



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#### System properties

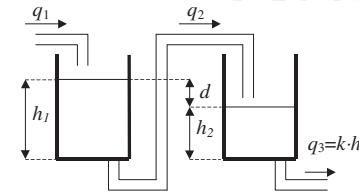
Dynamical v  
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## Linear v non-linear (cont.)

### Example

Consider a system consisting of two cylindric tanks, both of base  $B$  [m<sup>2</sup>]

- Output flow-rate from tank 1 is the input flow-rate to tank 2,  $q_2$



First tank

- Input flow-rate  $q_1$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_1$  is the liquid level [m]

Second tank

- Input flow-rate  $q_2$  [m<sup>3</sup>s<sup>-1</sup>]
- Output flow-rate  $q_3$  [m<sup>3</sup>s<sup>-1</sup>]
- $h_2$  is the liquid level [m]

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## Linear v non-linear (cont.)

The SS representation of the system

$$\rightsquigarrow \begin{cases} \rightsquigarrow \begin{cases} \frac{dV_1(t)}{dt} = q_1(t) - q_2(t) \\ \frac{dV_2(t)}{dt} = q_2(t) - q_3(t) = q_2(t) - kh_2(t) \end{cases} \\ \rightsquigarrow y(t) = \frac{1}{B}x_1(t) - \frac{1}{B}x_2(t) \end{cases}$$

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## Examples (cont.)

We obtained,

$$\rightsquigarrow \begin{cases} \dot{x}_1(t) &= u_1(t) - u_2(t) \\ \dot{x}_2(t) &= -\frac{k}{B}x_2(t) + u_2(t) \\ \dot{y}(t) &= \frac{1}{B}x_1(t) - \frac{1}{B}x_2(t) \end{cases}$$

We have,

$$\mathbf{A}(t) = \begin{bmatrix} 0 & 0 \\ 0 & -\frac{k}{B} \end{bmatrix}, \quad \mathbf{B}(t) = \begin{bmatrix} 1 & -1 \\ 0 & 1 \end{bmatrix}$$

$$\mathbf{C}(t) = \begin{bmatrix} \frac{1}{B} & -\frac{1}{B} \end{bmatrix}, \quad \mathbf{D}(t) = \begin{bmatrix} 0 & 0 \end{bmatrix}$$



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## Linear v non-linear (cont.)

### Example

#### Counter-intuition

Consider the system described by the IO model

$$y(t) = u(t) + 1$$

The system violates the superposition principle

- It is thus nonlinear

Consider two constant inputs

- $u_1(t) = 1$
- $u_2(t) = 2$

We can calculate the outputs

- $y_1(t) = u_1(t) + 1 = 2$
- $y_2(t) = u_2(t) + 2 = 3$

## Linear v non-linear (cont.)

Consider the combined input

$$u_3(t) = u_1(t) + u_2(t) = 3$$

The resulting output

$$y_3(t) = u_3(t) + 1 = 4$$

We thus have,

$$y_3(t) = 4 \neq y_1(t) + y_2(t) = 5$$

The IO representation is a nonlinear algebraic equation

- Blame the +1 on the RHS for nonlinearity

■

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## Linear v non-linear (cont.)

### Example

#### Counter-intuition

Consider the system described by the IO model

$$\dot{y}(t) + y(t) = \sqrt{t-1}u(t)$$

The system is linear

Consider the IO representation for a linear SISO system

$$\begin{aligned} a_0(t)y(t) + a_1(t)\dot{y}(t) + \dots + a_n(t)y^{(n)}(t) \\ = b_0(t)u(t) + b_1(t)\dot{u}(t) + \dots + b_m(t)u^{(m)}(t) \end{aligned}$$

$$\rightsquigarrow a_0(t) = 1$$

$$\rightsquigarrow a_1(t) = 1$$

$$\rightsquigarrow b_0(t) = \sqrt{t-1}$$

System  $\dot{y}(t) + y(t) = \sqrt{t-1}u(t)$  is thus linear

■

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## Stationary and non-stationary systems

### Definition

#### Stationarity

A system is said to be **stationary** (or **time invariant**), if it obeys the translation principle

A system is said to be **non-stationary** (or **time varying**), otherwise

## Stationary and non-stationary systems (cont.)

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### Translation principle

Consider some system

- Let the system response to a cause  $c_1(t)$  be equal to an effect  $e_1(t)$

System response to cause  $c_2(t) = c_1(t - T)$  equals effect  $e_2(t) = e_1(t - T)$

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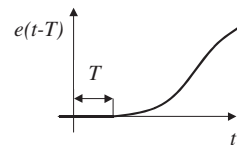
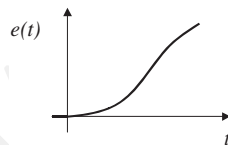
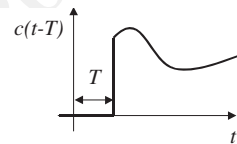
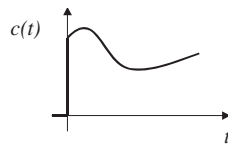
Proper v improper  
With v without  
delay

## Stationary and non-stationary systems (cont.)

Let the same cause be applied to system  $S$  at 2 different times

$\rightsquigarrow t = 0$

$\rightsquigarrow t = T$



The resulting effect is analogous

- Shifted by time interval  $T$

## Stationary and non-stationary systems (cont.)

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In nature no system is stationary

Yet, there exists a wide range of variations that can be neglected

- Over large time intervals

Over such intervals, the systems can be considered as stationary



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## Stationary and non-stationary systems (cont.)

### Proposition

#### IO representation

A necessary and sufficient condition for a system to be stationary

↪ The IO representation must not explicitly depend on time

Consider the SISO system

$$h[y(t), \dot{y}(t), \dots, y^{(n)}(t), u(t), \dot{u}(t), \dots, u^{(m)}(t), t] = 0$$

Then, the stationary model becomes

$$\rightsquigarrow h[y(t), \dot{y}(t), \dots, y^{(n)}(t), u(t), \dot{u}(t), \dots, u^{(m)}(t)] = 0$$

## Stationary and non-stationary systems (cont.)

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Consider a linear SISO system

$$\begin{aligned} a_0(t)y(t) + a_1(t)\dot{y}(t) + \dots + a_n(t)y^{(n)}(t) \\ = b_0(t)u(t) + b_1(t)\dot{u}(t) + \dots + b_m(t)u^{(m)}(t) \end{aligned}$$

The model becomes a linear differential equation

$$\rightsquigarrow a_0 y(t) + a_1 \dot{y}(t) + \dots + a_n y^{(n)}(t) = b_0 u(t) + b_1 \dot{u}(t) + \dots + b_m u^{(m)}(t)$$

The coefficients are constant

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## Stationary and non-stationary systems (cont.)

### Theorem

#### SS representation

A necessary and sufficient condition for a system to be stationary

↪ The SS representation must not explicitly depend on time

- (Both state equation and output transformation)

Consider the system

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), \mathbf{u}(t), t] \\ \mathbf{y}(t) = \mathbf{g}[\mathbf{x}(t), \mathbf{u}(t), t] \end{cases}$$

Then, the stationary model becomes

$$\rightsquigarrow \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), \mathbf{u}(t)] \\ \mathbf{y}(t) = \mathbf{g}[\mathbf{x}(t), \mathbf{u}(t)] \end{cases}$$

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## Stationary and non-stationary systems (cont.)

Consider a linear system

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}(t)\mathbf{x}(t) + \mathbf{B}(t)\mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C}(t)\mathbf{x}(t) + \mathbf{D}(t)\mathbf{u}(t) \end{cases}$$

The model becomes

$$\rightsquigarrow \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t) \end{cases}$$

The (elements of the) coefficient matrices **A**, **B**, **C** and **D** are constant

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## Stationary and non-stationary systems (cont.)

### Example

Consider the instantaneous and linear system

$$y(t) = tu(t)$$

The system is non-stationary

We can show this by using the translation principle

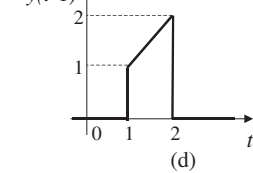
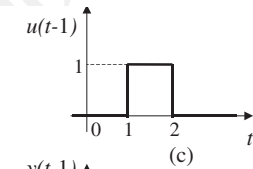
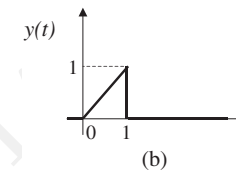
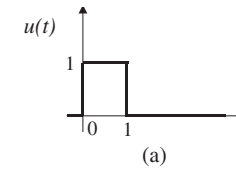
Consider the input

$$u(t) = \begin{cases} 1, & \text{if } t \in [0, 1] \\ 0, & \text{otherwise} \end{cases}$$

If the same input is applied with a delay, the output is not simply shifted

## Stationary and non-stationary systems (cont.)

The same input is applied with one (1) time-unit delay



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## Proper v improper Properties

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## Proper and improper systems

### Definition

#### Appropriateness

A system is said to be **proper**, if it obeys the causality principle

The system is said to be **improper**, otherwise

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## Proper and improper systems (cont.)

### Causality principle

The effect does not precede the generating cause

In nature, all systems are (obviously?) proper

- Only the model can be improper

## Proper v improper (cont.)

### Proposition

#### IO representation - SISO

A necessary and sufficient condition for a SISO system to be proper

↪ The order of derivation of the output ( $n$ ) is equal to or larger than the order of derivation of the input ( $m$ )

$$h[y(t), \dot{y}(t), \dots, y^{(n)}(t), u(t), \dot{u}(t), \dots, u^{(m)}(t), t] = 0, \text{ with } n \geq m$$

A system where  $n > m$  is said to be **strictly proper**

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## Proper and improper systems (cont.)

The result can be extended to MIMO systems

$$\begin{cases} h_1 \left[ \underbrace{y_1(t), \dot{y}_1(t), \dots, y_1^{(n_1)}(t)}_{\text{output 1}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{1,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{1,r})}(t)}_{\text{input r}} \right] \\ = 0 \\ h_2 \left[ \underbrace{y_2(t), \dot{y}_2(t), \dots, y_2^{(n_2)}(t)}_{\text{output 2}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{2,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{2,r})}(t)}_{\text{input r}} \right] \\ = 0 \\ \vdots \\ h_p \left[ \underbrace{y_p(t), \dot{y}_p(t), \dots, y_p^{(n_p)}(t)}_{\text{output p}}, \underbrace{u_1(t), \dot{u}_1(t), \dots, u_1^{(m_{p,1})}(t)}_{\text{input 1}}, \dots, \underbrace{u_r(t), \dots, u_r^{(m_{p,r})}(t)}_{\text{input r}} \right] \\ = 0 \end{cases}$$

None of the equations must include derivatives of the input variables whose order is larger than the derivation order of corresponding output variables

$$n_i \geq \max_{j=1, \dots, r} m_{i,j}, \quad \text{for all } i = 1, \dots, p$$

A system is strictly proper if the inequality is strictly true, for all  $i = 1, \dots, p$

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2018.1

### General concepts

Modelling,  
identification and  
analysis

Control,  
optimisation and  
validation

Fault diagnosis

### Classification

### Representation

Input-output  
representation

State-space  
representation

The model

### Examples

### System properties

Dynamical v  
Instantaneous

Linear v Nonlinear

Stationary v  
non-stationary

Proper v improper

With v without  
delay

## Proper and improper systems (cont.)

### Proposition

#### SS representation

Consider a system described by a SS model

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), \mathbf{u}(t), t] \\ \mathbf{y}(t) = \mathbf{g}[\mathbf{x}(t), \mathbf{u}(t), t] \end{cases}$$

Such a system/model is always proper

A strictly proper system has an output transformation independent on  $\mathbf{u}(t)$

$$\rightsquigarrow \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}[\mathbf{x}(t), \mathbf{u}(t), t] \\ \mathbf{y}(t) = \mathbf{g}[\mathbf{x}(t), t] \end{cases}$$

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## Proper and improper systems (cont.)

The SS model of a linear, stationary and strictly proper system

$$\rightsquigarrow \begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t) \\ \mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) \end{cases}$$

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**With v without delay**  
**Properties**

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## Systems with and without delay

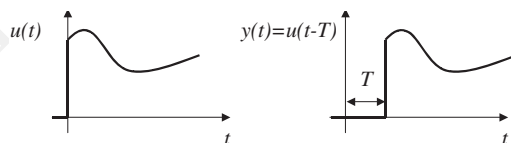
### Definition

#### **Finite time delay**

A **finite delay** is a system whose output  $y(t)$  at time  $t$  is equal to the input  $u(t - T)$  at time  $t - T$



$\rightsquigarrow T \in (0, +\infty)$  is called the **time delay**



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## Systems with and without delay (cont.)

Consider the algebraic equation describing a finite delay element

$$y(t) = u(t - T), \text{ with } T \in \mathcal{R}^+$$

- Such a system is not instantaneous
- The system is dynamic

The output at time  $t$  depends on the previous values of the input

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## Systems with and without delay (cont.)

### Proposition

#### *IO and SS representation*

*A necessary and sufficient condition for a system to be without a time delay*

*↔ All the signals in the model (IO or SS) must share the same argument*



## Systems with and without delay (cont.)

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### Example

Consider a system described by the IO model

$$4\dot{y}(t) + 2y(t) = u(t - T)$$

The system has delay elements

- There are signals that are dependent on  $t$
- There are signals that are dependent on  $t - T$



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## Systems with and without delay (cont.)

### Example

Consider a system described by the SS model

$$\begin{cases} \dot{x}(t) = x(t - T) + u(t) \\ y(t) = 7x(t) \end{cases}$$

The system/model has delay elements

- There are signals that are dependent on  $t$
- There are signals that are dependent on  $t - T$

